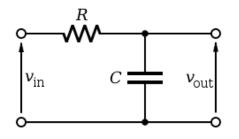
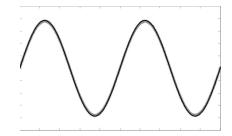
# EECS 16

Logo credits go to Moses Won

## Discussion 5B

**Transfer Functions & Filters** 







### Recap

If the voltages and currents in a linear circuit are **sinusoidal**, then we can use **phasors** to analyze our circuit.

• A linear circuit **does not** change the frequency of a sinusoid.

Let's say the voltage  $v(t) = A \cos(\omega t + \phi)$ , then the phasor  $V = Ae^{j\phi}$ .

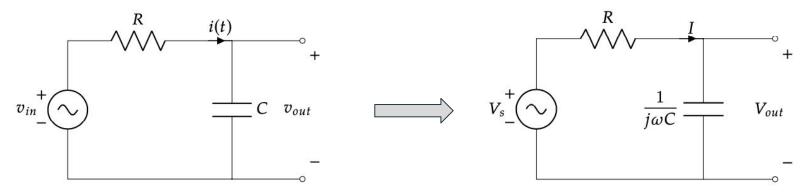
We could transform our entire circuit into the phasor domain using impedances!

- Resistors have impedance R
- Capacitors have impedance  $1/j\omega C$
- Inductors have impedance jωL

### Quick Example

Given the following circuit in the time-domain

We can transform it into the phasor domain



Then, we can solve for phasor voltage  $V_{out}$  and convert that back into the time domain.

Ex: If  $V_{out} = 2e^{j \pi/4}$ , then  $v_{out}(t) = 2 \cos(\omega t + \pi/4)$ .

### Plan for Today

We want to see how our circuits behave for inputs of different frequencies.

Let's define a **transfer function** which will be output to input ratio of the phasors voltages:  $H(\omega) = V_{out} / V_{in}$ 

- The transfer function takes frequencies  $\omega$  and outputs a complex scalar.
- It's the frequency equivalent of **gain** from 16A:  $G = V_{out} / V_{in}$

Transfer functions can help us design filters that block out certain frequencies.

Many applications such as audio-engineering, WiFi, computer graphics.

#### Poles and Zeros

Any transfer function can be factored into the following form:

$$H(\omega) = \frac{z(\omega)}{p(\omega)} = \frac{(j\omega)^{N_{z0}}}{(j\omega)^{N_{p0}}} \left( \frac{(j\omega)^{n}\alpha_{n} + (j\omega)^{n-1}\alpha_{n-1} + \dots + j\omega\alpha_{1} + \alpha_{0}}{(j\omega)^{m}\beta_{m} + (j\omega)^{m-1}\beta_{m-1} + \dots + j\omega\beta_{1} + \beta_{0}} \right)$$

$$= K \frac{(j\omega)^{N_{z0}} \left( 1 + j\frac{\omega}{\omega_{z1}} \right) \left( 1 + j\frac{\omega}{\omega_{z2}} \right) \dots \left( 1 + j\frac{\omega}{\omega_{zn}} \right)}{(j\omega)^{N_{p0}} \left( 1 + j\frac{\omega}{\omega_{p1}} \right) \left( 1 + j\frac{\omega}{\omega_{p2}} \right) \dots \left( 1 + j\frac{\omega}{\omega_{pm}} \right)}$$

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$$= K \frac{(j\omega)^{N_{z0}} \left( 1 + j\frac{\omega}{\omega_{z1}} \right) \left( 1 + j\frac{\omega}{\omega_{z2}} \right) \dots \left( 1 + j\frac{\omega}{\omega_{zm}} \right)}{(j\omega)^{N_{p0}} \left( 1 + j\frac{\omega}{\omega_{p1}} \right) \left( 1 + j\frac{\omega}{\omega_{p2}} \right) \dots \left( 1 + j\frac{\omega}{\omega_{pm}} \right)}$$

The  $\omega_z$  are called zeros and the  $\omega_p$  are called poles. Really important for plotting

**Ex:**  $H_2(\omega) = \frac{\Im \omega (1 + j\omega/50)}{(1 + j\omega/100)(1 + j\omega/5)}$ 

plotting
Where things start to change.

This transfer function has a zero at 50, and poles at 5, 100, also have 9

Zero at 
$$W=0$$
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